

LOGIC SYNTHESIS

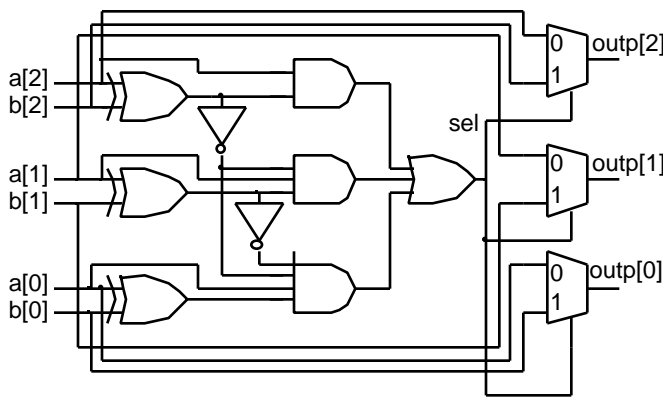
12

Key terms and concepts: **logic synthesis** converts an HDL **behavioral model** (Verilog or VHDL) to a netlist (**structural model**) the same way a C compiler converts C code to machine language • a cell library is called the **target library**

12.1 A Logic-Synthesis Example

A comparison of hand design with synthesis (using a 1.0 μm VLSI Technology cell library)

	Path delay/ ns	No. of standard cells	No. of transistors	Chip area/ mils ²
Hand design	41.6	1,359	16,545	21,877
Synthesized design	36.3	1,493	11,946	18,322



```
// comp_mux.v
module comp_mux(a, b, outp);
  input [2:0] a, b;
  output [2:0] outp;
  function [2:0] compare;
    input [2:0] ina, inb;
  begin
    if (ina <= inb) compare = ina;
    else compare = inb;
  end
endfunction
assign outp = compare(a, b);
endmodule
```

Comparison of the comparator/MUX designs using a 1.0 μm standard-cell library

	Delay /ns	No. of standard cells	No. of transistors	Area /mils ²
Hand design	4.3	12	116	68.68
Synthesized	2.9	15	66	46.43

12.2 A Comparator/MUX

Key terms and concepts: synopsys_dc.setup • script • derived schematic • analysis • elaboration • logic optimization • logic-mapping • timing-analysis (timing engine)

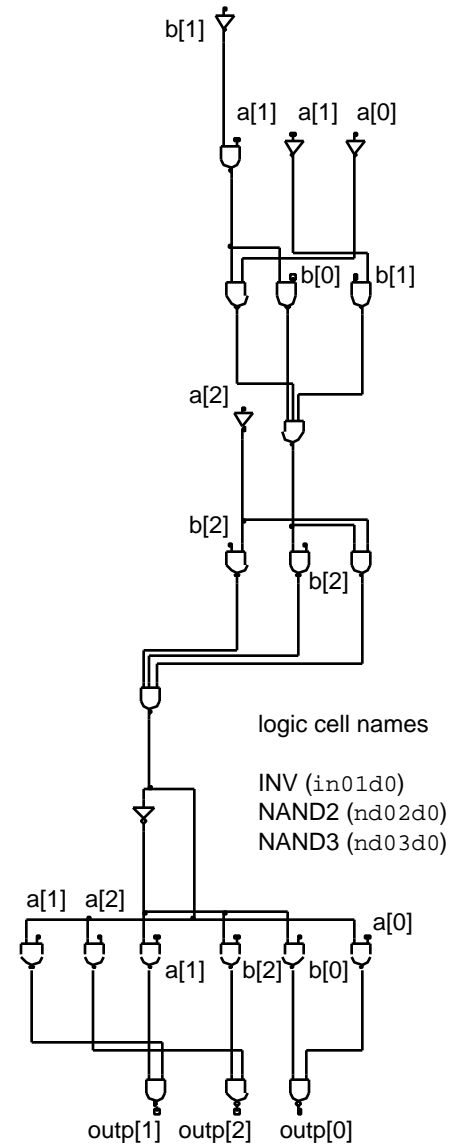
```

`timescale 1ns / 10ps
module comp_mux_u (a, b, outp);
input  [2:0] a; input  [2:0] b;
output [2:0] outp;
supply1 VDD; supply0 VSS;

in01d0 u2 (.I(b[1]), .ZN(u2_ZN));
nd02d0 u3 (.A1(a[1]), .A2(u2_ZN), .ZN(u3_ZN));
in01d0 u4 (.I(a[1]), .ZN(u4_ZN));
nd02d0 u5 (.A1(u4_ZN), .A2(b[1]), .ZN(u5_ZN));
in01d0 u6 (.I(a[0]), .ZN(u6_ZN));
nd02d0 u7 (.A1(u6_ZN), .A2(u3_ZN), .ZN(u7_ZN));
nd02d0 u8 (.A1(b[0]), .A2(u3_ZN), .ZN(u8_ZN));
nd03d0 u9 (.A1(u5_ZN), .A2(u7_ZN), .A3(u8_ZN),
.ZN(u9_ZN));
in01d0 u10 (.I(a[2]), .ZN(u10_ZN));
nd02d0 u11 (.A1(u10_ZN), .A2(u9_ZN),
.ZN(u11_ZN));
nd02d0 u12 (.A1(b[2]), .A2(u9_ZN), .ZN(u12_ZN));
nd02d0 u13 (.A1(u10_ZN), .A2(b[2]), .ZN(u13_ZN));
nd03d0 u14 (.A1(u11_ZN), .A2(u12_ZN),
.A3(u13_ZN), .ZN(u14_ZN));
nd02d0 u15 (.A1(a[2]), .A2(u14_ZN), .ZN(u15_ZN));
in01d0 u16 (.I(u14_ZN), .ZN(u16_ZN));
nd02d0 u17 (.A1(b[2]), .A2(u16_ZN), .ZN(u17_ZN));
nd02d0 u18 (.A1(u15_ZN), .A2(u17_ZN),
.ZN(outp[2]));
nd02d0 u19 (.A1(a[1]), .A2(u14_ZN), .ZN(u19_ZN));
nd02d0 u20 (.A1(b[1]), .A2(u16_ZN), .ZN(u20_ZN));
nd02d0 u21 (.A1(u19_ZN), .A2(u20_ZN),
.ZN(outp[1]));
nd02d0 u22 (.A1(a[0]), .A2(u14_ZN), .ZN(u22_ZN));
nd02d0 u23 (.A1(b[0]), .A2(u16_ZN), .ZN(u23_ZN));
nd02d0 u24 (.A1(u22_ZN), .A2(u23_ZN),
.ZN(outp[0]));

endmodule

```



The comparator/MUX after logic synthesis, but before logic optimization

The structural netlist, `comp_mux_u.v`, and its derived schematic

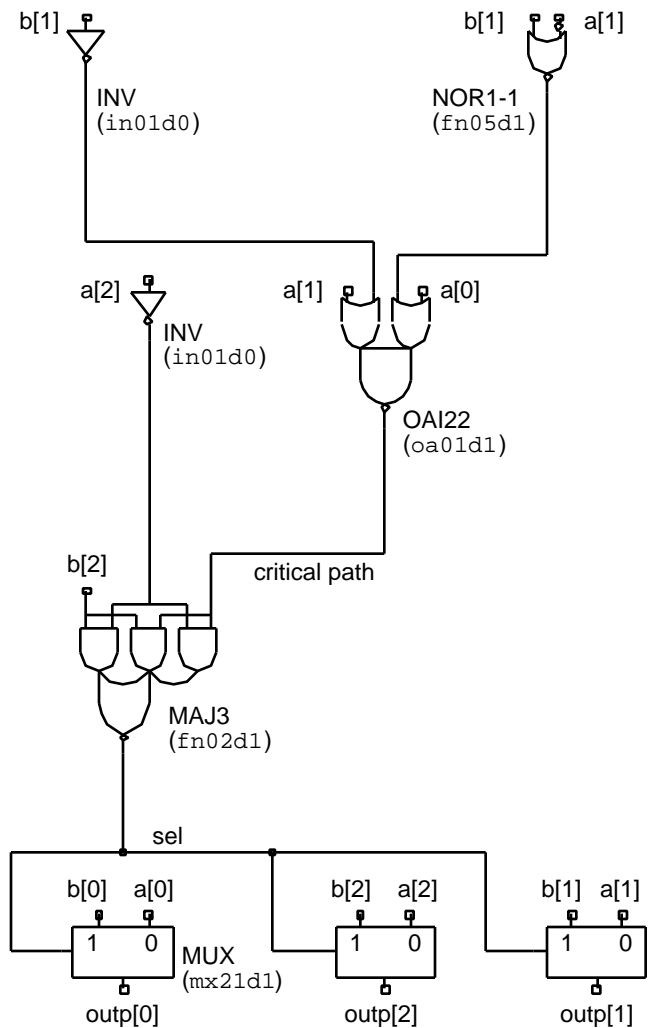
```

`timescale 1ns / 10ps
module comp_mux_o (a, b, outp);
input [2:0] a; input [2:0] b;
output [2:0] outp;
supply1 VDD; supply0 VSS;

in01d0 B1_i1 (.I(a[2]),
.ZN(B1_i1_ZN));
in01d0 B1_i2 (.I(b[1]),
.ZN(B1_i2_ZN));
oa01d1 B1_i3 (.A1(a[0]),
.A2(B1_i4_ZN), .B1(B1_i2_ZN),
.B2(a[1]), .ZN(B1_i3_Z);
fn05d1 B1_i4 (.A1(a[1]), .B1(b[1]),
.ZN(B1_i4_ZN));
fn02d1 B1_i5 (.A(B1_i3_ZN),
.B(B1_i1_ZN), .C(b[2]),
.ZN(B1_i5_ZN));
mx21d1 B1_i6 (.I0(a[0]), .I1(b[0]),
.S(B1_i5_ZN), .Z(outp[0]));
mx21d1 B1_i7 (.I0(a[1]), .I1(b[1]),
.S(B1_i5_ZN), .Z(outp[1]));
mx21d1 B1_i8 (.I0(a[2]), .I1(b[2]),
.S(B1_i5_ZN), .Z(outp[2]));

endmodule

```



The comparator/MUX after logic synthesis and logic optimization with the default settings

The structural netlist, `comp_mux_o.v`, and its derived schematic

12.2.1 An Actel Version of the Comparator/MUX

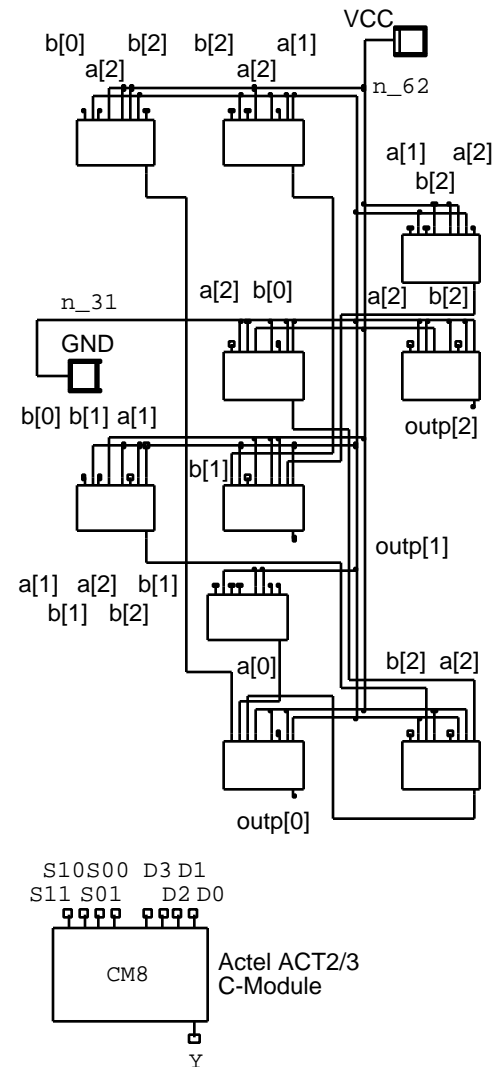
Key terms and concepts: Actel ACT 2/3 FPGA architecture • the symbols represent the eight-input ACT 2/3 C-Module • the logic synthesizer, in the technology-mapping step, decides the connections to the inputs to the logic macro, CM8

```

`timescale 1 ns/100 ps
module comp_mux_actel_o (a, b, outp);
input [2:0] a, b; output [2:0] outp;
wire n_13, n_17, n_19, n_21, n_23, n_27, n_29,
n_31, n_62;

CM8 I_5_CM8(.D0(n_31), .D1(n_62), .D2(a[0]),
.D3(n_62), .S00(n_62), .S01(n_13), .S10(n_23),
.S11(n_21), .Y(outp[0]));
CM8 I_2_CM8(.D0(n_31), .D1(n_19), .D2(n_62),
.D3(n_62), .S00(n_62), .S01(b[1]), .S10(n_31),
.S11(n_17), .Y(outp[1]));
CM8 I_1_CM8(.D0(n_31), .D1(n_31), .D2(b[2]),
.D3(n_31), .S00(n_62), .S01(n_31), .S10(n_31),
.S11(a[2]), .Y(outp[2]));
VCC VCC_I(.Y(n_62));
CM8 I_4_CM8(.D0(a[2]), .D1(n_31), .D2(n_62),
.D3(n_62), .S00(n_62), .S01(b[2]), .S10(n_31),
.S11(a[1]), .Y(n_19));
CM8 I_7_CM8(.D0(b[1]), .D1(b[2]), .D2(n_31),
.D3(n_31), .S00(a[2]), .S01(b[1]), .S10(n_31),
.S11(a[1]), .Y(n_23));
CM8 I_9_CM8(.D0(n_31), .D1(n_31), .D2(a[1]),
.D3(n_31), .S00(n_62), .S01(b[1]), .S10(n_31),
.S11(b[0]), .Y(n_27));
CM8 I_8_CM8(.D0(n_29), .D1(n_62), .D2(n_31),
.D3(a[2]), .S00(n_62), .S01(n_27), .S10(n_31),
.S11(b[2]), .Y(n_13));
CM8 I_3_CM8(.D0(n_31), .D1(n_31), .D2(a[1]),
.D3(n_31), .S00(n_62), .S01(a[2]), .S10(n_31),
.S11(b[2]), .Y(n_17));
CM8 I_6_CM8(.D0(b[2]), .D1(n_31), .D2(n_62),
.D3(n_62), .S00(n_62), .S01(a[2]), .S10(n_31),
.S11(b[0]), .Y(n_21));
CM8 I_10_CM8(.D0(n_31), .D1(n_31), .D2(b[0]),
.D3(n_31), .S00(n_62), .S01(n_31), .S10(n_31),
.S11(a[2]), .Y(n_29));
GND GND_I(.Y(n_31));
endmodule

```



The Actel version of the comparator/MUX after logic optimization

The structural netlist, `comp_mux_actel_o_ad1_e.y` and its derived schematic

12.3 Inside a Logic Synthesizer

Key terms and concepts: The logic synthesizer parses the Verilog and builds an internal data structure (CDFG) • **logic minimization** finds a minimum **cover** • **synthesized network** • **logic optimization** uses factoring, substitution, and elimination • **technology-decomposition** builds a generic network • **technology-mapping (logic-mapping)** matches pieces of the network with the logic cells • we **imply** A • the logic synthesizer has to **infer** B • we must write HDL code so A=B

12.4 Synthesis of the Viterbi Decoder

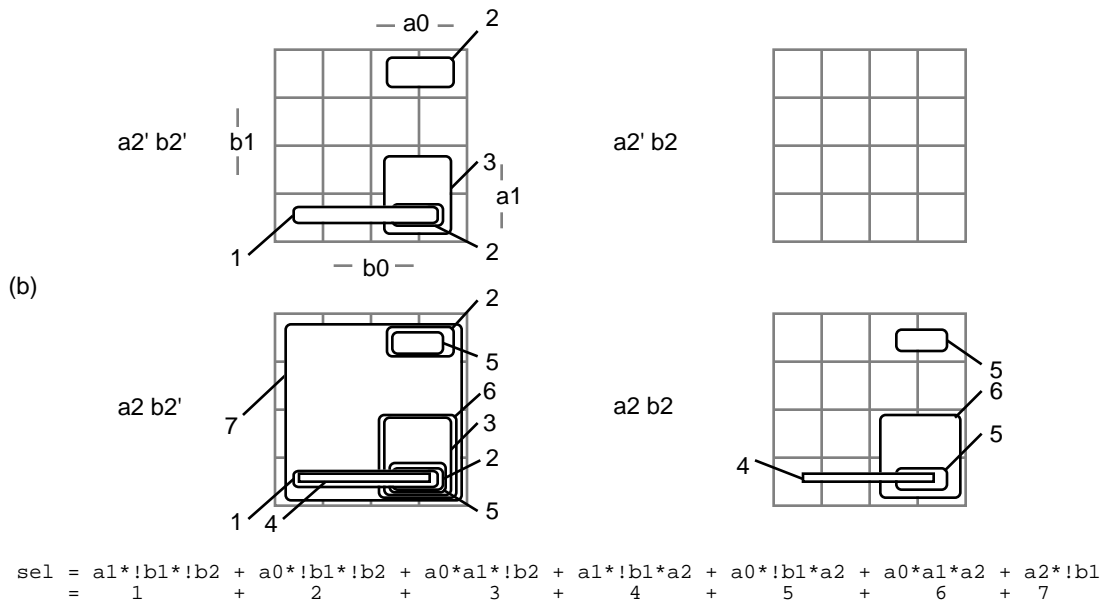
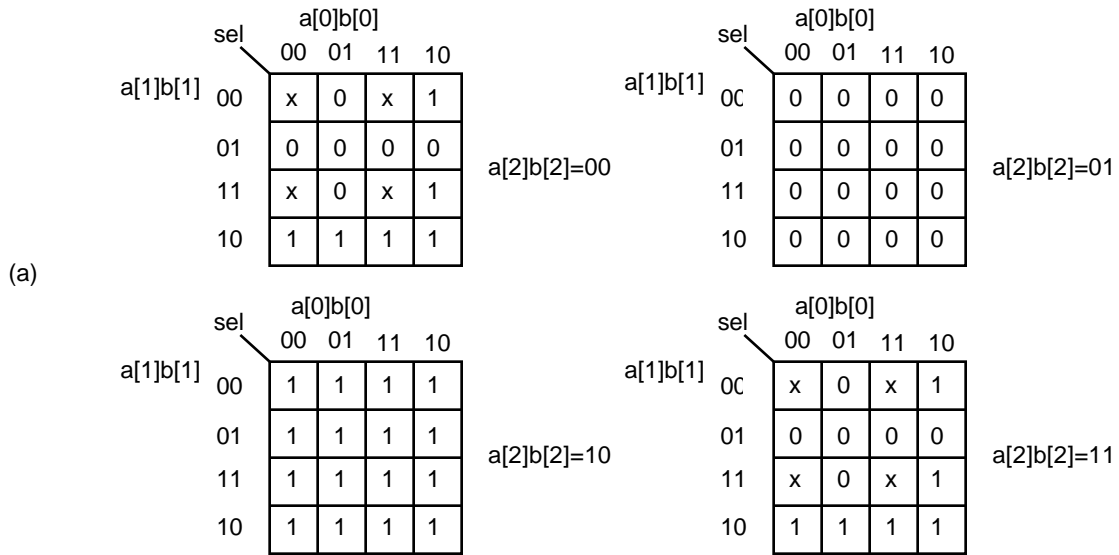
12.4.1 ASIC I/O

Key terms and concepts: inference of I/O cells • directives for special pads (clock buffers) • pull-up resistor, slew rate • no standards • no accepted way to set these parameters from an HDL • generic technology-independent I/O models • instantiate I/O cells directly from a library

```
// asPadBidir #(W, N, S, L, P) I (Pad, toCore, frCore, OEN) //1
// W = width, integer (default=1) //2
// N = pin number string, e.g. "1:3,5:8" //3
// S = strength = {2, 4, 8, 16} in mA drive //4
// L = level = {cmos, ttl, schmitt} (default = cmos) //5
// P = pull-up resistor = {down, float, none, up} //6
// Vxx = {Vss, Vdd} //7

module PadTri (Pad, I, Oen); // active-low output enable //1
parameter width = 1, pinNumbers = "", \strength = 1, //2
    level = "CMOS", externalVdd = 5; //3
output [width-1:0] Pad; input [width-1:0] I; input Oen; //4
assign #1 Pad = (Oen ? {width{1'bz}} : I); //5
endmodule //6

module PadBidir (C, Pad, I, Oen); // active-low output enable //1
parameter width = 1, pinNumbers = "", \strength = 1, //2
    level = "CMOS", pull = "none", externalVdd = 5; //3
output [width-1:0] C; inout [width-1:0] Pad; //4
input [width-1:0] I; input Oen; //5
assign #1 Pad = Oen ? {width{1'bz}} : I; assign #1 C = Pad; //6
endmodule //7
```



Logic maps for the comparator/MUX

(a) If the input b is less than a, then sel is '1'. If a=b, then sel = 'x' (don't care)

(b) A cover for sel.

12.4.2 Flip-Flops

Key terms and concepts: synthesis tools cannot handle two wait statements

```

module dff(D,Q,Clock,Reset); // N.B. reset is active-low //1
output Q; input D,Clock,Reset; //2
parameter CARDINALITY = 1; reg [CARDINALITY-1:0] Q; //3
wire [CARDINALITY-1:0] D; //4
always @(posedge Clock) if (Reset!=0) #1 Q=D; //5
always begin wait (Reset==0); Q=0; wait (Reset==1); end //6
endmodule //7

module dff(D, Q, Clk, Rst); // new flip-flop for Viterbi decoder //1
  parameter width = 1, reset_value = 0; input [width - 1 : 0] D; //2
  output [width - 1 : 0] Q; reg [width - 1 : 0] Q; input Clk, Rst; //3
  initial Q <= {width{1'bx}}; //4
  always @ ( posedge Clk or negedge Rst ) //5
    if ( Rst == 0 ) Q <= #1 reset_value; else Q <= #1 D; //6
endmodule //7

```

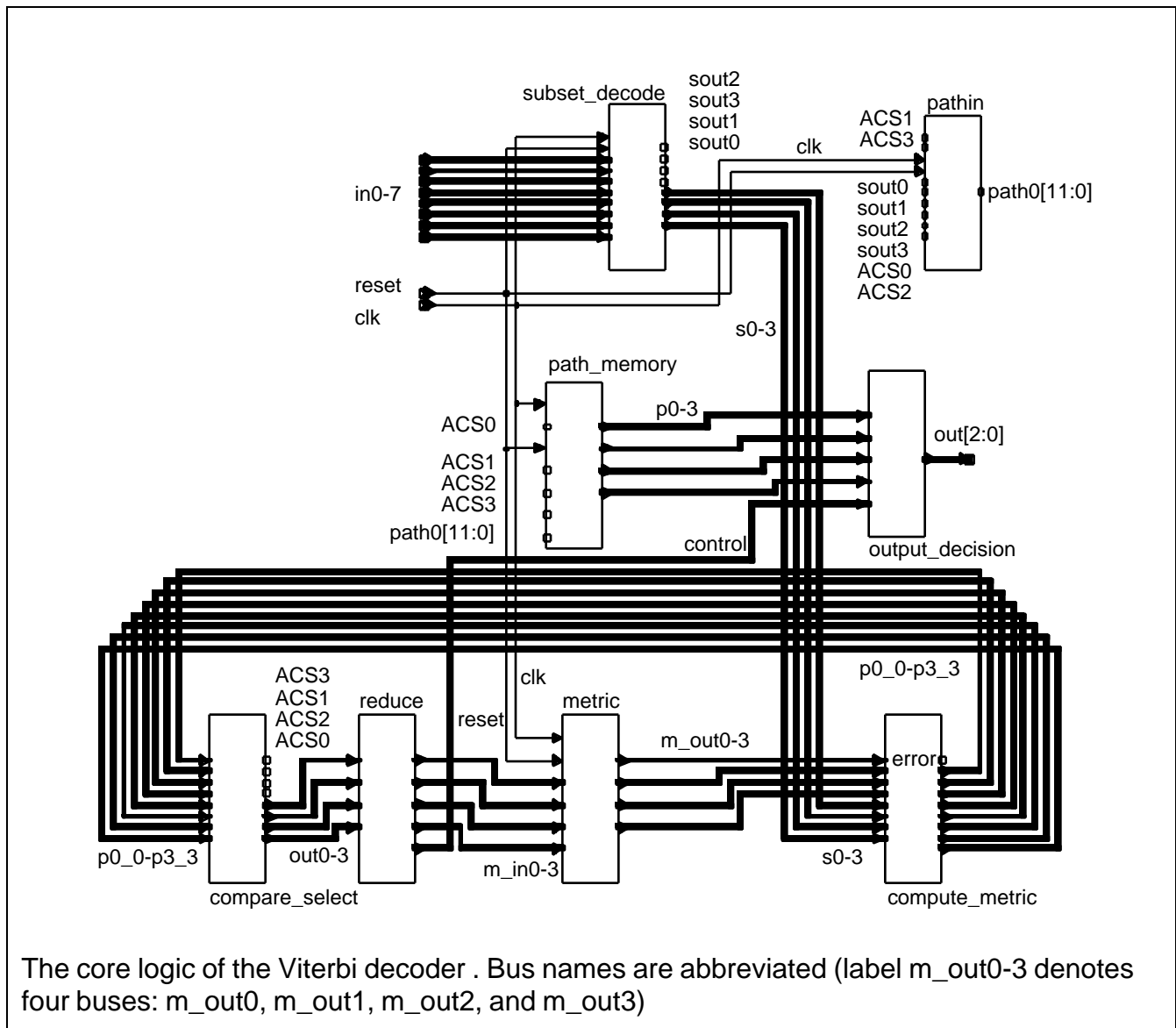
12.4.3 The Top-Level Model

Key terms and concepts: top-level Viterbi decoder • generic input, output, power, and clock I/O cells from the standard-component library

```

/* This is the top-level module, viterbi_ASIC.v */ //1
module viterbi_ASIC //2
(padin0, padin1, padin2, padin3, padin4, padin5, padin6, padin7, //3
padOut, padClk, padRes, padError); //4
input [2:0] padin0, padin1, padin2, padin3, //5
      padin4, padin5, padin6, padin7; //6
input padRes, padClk; output padError; output [2:0] padOut; //7
wire Error, Clk, Res; wire [2:0] Out; // core //8
wire padError, padClk, padRes; wire [2:0] padOut; //9
wire [2:0] in0,in1,in2,in3,in4,in5,in6,in7; // core //10
wire [2:0] //11
  padin0, padin1,padin2,padin3,padin4,padin5,padin6,padin7; //12
// Do not let the software mess with the pads. //13
//compass dontTouch u* //14
  asPadIn #(3,"1,2,3") u0 (in0, padin0); //15
  asPadIn #(3,"4,5,6") u1 (in1, padin1); //16
  asPadIn #(3,"7,8,9") u2 (in2, padin2); //17
  asPadIn #(3,"10,11,12") u3 (in3, padin3); //18
  asPadIn #(3,"13,14,15") u4 (in4, padin4); //19

```

```

asPadIn #(3,"16,17,18") u5 (in5, padin5); //20
asPadIn #(3,"19,20,21") u6 (in6, padin6); //21
asPadIn #(3,"22,23,24") u7 (in7, padin7); //22
asPadVdd #("25","both") u25 (vddb); //23
asPadVss #("26","both") u26 (vssb); //24
asPadClk #("27") u27 (Clk, padClk); //25
asPadOut #(1,"28") u28 (padError, Error); //26
asPadin #(1,"29") u29 (Res, padRes); //27
asPadOut #(3,"30,31,32") u30 (padOut, Out); //28
// Here is the core module: //29
viterbi v_1 //30

```

```
(in0,in1,in2,in3,in4,in5,in6,in7,Out,Clk,Res,Error); //31  
endmodule //32
```

12.5 Verilog and Logic Synthesis

Key terms and concepts: top-down design approach • **stubs** contain a minimum of code

```

module MyChip_ASIC()
  // behavioral "always", etc. ...
  SecondLevelStub1 port mapping
  SecondLevelStub2 port mapping
  ... endmodule
module SecondLevelStub1() ... assign Output1 = ~Input1; endmodule
module SecondLevelStub2() ... assign Output2 = ~Input2;
endmodule

```

12.5.1 Verilog Modeling

Key terms and concepts: **synthesizable** • **synthesis policy** • **modeling style** • **functionally identical**, or **functionally equivalent**

12.5.2 Delays in Verilog

Key terms and concepts: Synthesis tools ignore delay values

```

module Step_Time(clk, phase); //1
  input clk; output [2:0] phase; reg [2:0] phase; //2
  always @(posedge clk) begin //3
    phase <= 4'b0000; //4
    phase <= #1 4'b0001; phase <= #2 4'b0010; //5
    phase <= #3 4'b0011; phase <= #4 4'b0100; //6
  end //7
endmodule //8

module Step_Count (clk_5x, phase); //1
  input clk_5x; output [2:0] phase; reg [2:0] phase; //2
  always@(posedge clk_5x) //3
  case (phase) //4
    0:phase = #1 1; 1:phase = #1 2; 2:phase = #1 3; 3:phase = #1 4; //5
    default: phase = #1 0; //6
  endcase //7
endmodule //8

```

12.5.3 Blocking and Nonblocking Assignments

Key terms and concepts: **race condition** (or a **race**)

```

module race(clk, q0); input clk, q0; reg q1, q2;
always @(posedge clk) q1 = #1 q0; always @(posedge clk) q2 = #1 q1;
endmodule

```

```

module no_race_1(clk, q0, q2); input clk, q0; output q2; reg q1, q2;
always @(posedge clk) begin q2 = q1; q1 = q0; end
endmodule

```

```

module no_race_2(clk, q0, q2); input clk, q0; output q2; reg q1, q2;
always @(posedge clk) q1 <= #1 q0; always @(posedge clk) q2 <= #1 q1;
endmodule

```

12.5.4 Combinational Logic in Verilog

Key terms and concepts: level-sensitive sensitivity list • continuous assignment statements also imply combinational logic

```

module And_Always(x, y, z); input x,y; output z; reg z;
  always @(x or y) z <= x & y; // combinational logic method 1
endmodule

```

```

module And_Assign(x, y, z); input x,y; output z; wire z;
  assign z <= x & y; // combinational logic method 2 = method 1
endmodule

```

```

module And_Or (a,b,c,z); input a,b,c; output z; reg [1:0]z;
  always @(a or b or c) begin z[1]<= &{a,b,c}; z[2]<= |{a,b,c};end
endmodule

```

```

module Parity (BusIn, outp); input[7:0] BusIn; output outp; reg outp;
  always @(BusIn) if (^BusIn == 0) outp = 1; else outp = 0;
endmodule

```

```

module And_Bad(a, b, c); input a, b; output c; reg c;
always@(a) c <= a & b; // b is missing from this sensitivity list
endmodule

```

```

module CL_good(a, b, c); input a, b; output c; reg c;
always@(a or b)
begin c = a + b; d = a & b; e = c + d;end // c, d: LHS before RHS
endmodule

```

```

module CL_bad(a, b, c); input a, b; output c; reg c;
always@(a or b)
begin e = c + d; c = a + b; d = a & b;end // c, d: RHS before LHS
endmodule

```

```

// The complement of this function is too big for synthesis.
module Achilles (out, in); output out; input [30:1] in;
assign out = in[30]&in[29]&in[28] | in[27]&in[26]&in[25]
           | in[24]&in[23]&in[22] | in[21]&in[20]&in[19]
           | in[18]&in[17]&in[16] | in[15]&in[14]&in[13]
           | in[12]&in[11]&in[10] | in[9]& in[8]&in[7]
           | in[6] & in[5]&in[4] | in[3] & in[2]&in[1];
endmodule

```

12.5.5 Multiplexers In Verilog

Key terms and concepts: We imply a MUX using a case or if statement • metalogical values or **simbits** (such as 'x') are not “real” • avoid using casex and casez statements • if you need to “remember” a value, this implies sequential logic

```

module Mux_21a(sel, a, b, z); input sel, a , b; output z; reg z;
always @(a or b or sel)
begin case(sel) 1'b0: z <= a; 1'b1: z <= b;end
endmodule

```

```

module Mux_x(sel, a, b, z); input sel, a, b; output z; reg z;
always @(a or b or sel)
begin case(sel) 1'b0: z <= 0; 1'b1: z <= 1; 1'bx: z <= 'x';end
endmodule

```

```

module Mux_21b(sel, a, b, z); input sel, a, b; output z; reg z;
always @(a or b or sel) begin if (sel) z <= a else z <= b; end
endmodule

```

```

module Mux_Latch(sel, a, b, z); input sel, a, b; output z; reg z;
always @(a or sel) begin if (sel) z <= a; end
endmodule

```

```

module Mux_81(InBus, sel, OE, OutBit); //1
input [7:0] InBus; input [2:0] Sel; //2
input OE; output OutBit; reg OutBit; //3
always @(OE or sel or InBus) //4
  begin //5
    if (OE == 1) OutBit = InBus[sel]; else OutBit = 1'bz; //6
  end //7
endmodule //8

```

12.5.6 The Verilog Case Statement

Key terms and concepts: **exhaustive** • **compiler directive** • **synthesis directive** • **pseudocomment** • an 'x' (**synthesis don't care value**) gives the synthesizer flexibility in optimization • **priority encoder**

```

module case8_oneHot(oneHot, a, b, c, z); //1
input a, b, c; input [2:0] oneHot; output z; reg z; //2
always @(oneHot or a or b or c) //3
begin case(oneHot) //synopsys full_case //4
  3'b001: z <= a; 3'b010: z <= b; 3'b100: z <= c; //5
  default: z <= 1'bx; endcase //6
end //7
endmodule //8

```

```

module case8_priority(oneHot, a, b, c, z); //1
input a, b, c; input [2:0] oneHot; output z; reg z; //2
always @(oneHot or a or b or c) begin //3
case(1'b1) //synopsys parallel_case //4
  oneHot[0]: z <= a; //5
  oneHot[1]: z <= b; //6
  oneHot[2]: z <= c; //7
  default: z <= 1'bx; endcase //8
end //9
endmodule //10

```

12.5.7 Decoders In Verilog

Key terms and concepts: the synthesizer infers a three-state buffer from an assignment of 'z'

```

module Decoder_4To16(enable, In_4, Out_16); // 4-to-16 decoder //1
input enable; input [3:0] In_4; output [15:0] Out_16; //2
reg [15:0] Out_16; //3
always @(enable or In_4) //4
    begin Out_16 = 16'hzzzz; //5
        if (enable == 1) //6
            begin Out_16 = 16'h0000; Out_16[In_4] = 1; end //7
        end //8
    endmodule //9

if (enable === 1) // can't make logic to check for enable = x or z

```

12.5.8 Priority Encoder in Verilog

Key terms and concepts: The logic synthesizer must be able to unroll a loop in a for statement.

```

module Pri_Encoder32 (InBus, Clk, OE, OutBus); //1
input [31:0]InBus; input OE, Clk; output [4:0]OutBus; //2
reg j; reg [4:0]OutBus; //3
always@(posedge Clk) //4
    begin //5
        if (OE == 0) OutBus = 5'bz ; //6
    else //7
        begin OutBus = 0; //8
            for (j = 31; j >= 0; j = j - 1) //9
                begin if (InBus[j] == 1) OutBus = j; end //10
            end //11
        end //12
    endmodule //13

```

12.5.9 Arithmetic in Verilog

Key terms and concepts: make room for the carry bit when you add two numbers in Verilog •

resource allocation • resource sharing • multiplication assumes nets are unsigned

```

module Adder_8 (A, B, Z, Cin, Cout); //1
input [7:0] A, B; input Cin; output [7:0] Z; output Cout; //2
assign {Cout, Z} = A + B + Cin; //3
endmodule //4

module Adder_16 (A, B, Sum, Cout); //1
input [15:0] A, B; output [15:0] Sum; output Cout; //2

```

```

reg [15:0] Sum; reg Cout; //3
always @(A or B) {Cout, Sum} = A + B + 1; // One adder not two! //4
endmodule //5

module Add_A (sel, a, b, c, d, y); //1
input a, b, c, d, sel; output y; reg y; //2
always@(sel or a or b or c or d) // One or two adders? //3
begin if (sel == 0) y <= a + b; else y <= c + d; end //4
endmodule //5

module Add_B (sel, a, b, c, d, y); //1
input a, b, c, d, sel; output y; reg t1, t2, y; //2
always@(sel or a or b or c or d) begin // One adder not two! //3
if (sel == 0) begin t1 = a; t2 = b; end // Temporary //4
else begin t1 = c; t2 = d; end // variables. //5
y = t1 + t2; end //6
endmodule //7

module Multiply_unsigned (A, B, Z); //1
input [1:0] A, B; output [3:0] Z; //2
assign Z <= A * B; //3
endmodule //4

module Multiply_signed (A, B, Z); //1
input [1:0] A, B; output [3:0] Z; //2
// 00 -> 00_00 01 -> 00_01 10 -> 11_10 11 -> 11_11 //3
assign Z = { { 2{A[1]} }, A} * { { 2{B[1]} }, B}; //4
endmodule //5

```

12.5.10 Sequential Logic in Verilog

Key terms and concepts: edges (**posedge** or **negedge**) in the sensitivity list of an **always** statement imply a clocked storage element • however, an **always** statement does not have to be edge-sensitive to imply sequential logic • all sequential logic cells must be initialized • **template** • **synthesis style guide**

```

always@(posedge clock) Q_flipflop = D; // A flip-flop.
always@(clock or D) if (clock) Q_latch = D; // A latch.
always@(posedge clock or negedge reset) // names mean nothing,
always@(posedge day or negedge year) // which is the reset?

```



```

always@(posedge clk or negedge reset) begin // Template for reset:
    if (reset == 0) Q = 0; // initialize,
    else Q = D;           // normal clocking
end

```

```

module Counter_With_Reset (count, clock, reset); //1
input clock, reset; output count; reg [7:0] count; //2
always @ (posedge clock or negedge reset) //3
    if (reset == 0) count = 0; else count = count + 1; //4
endmodule //5

```

```

module DFF_MasterSlave (D, clock, reset, Q); // D type flip-flop //1
input D, clock, reset; output Q; reg Q, latch; //2
always @(posedge clock or posedge reset) //3
    if (reset == 1) latch = 0; else latch = D; // the master. //4
always @(latch) Q = latch; // the slave. //5
endmodule //6

```

12.5.11 Component Instantiation in Verilog

Key terms and concepts: HDL description is technology-independent (CMOS, FPGA, TTL, GaAs) • the only way to use a particular cell is to use structural Verilog and **hand instantiation**

- dont_touch • **soft models** or **standard components** • **DesignWare**

```

//Compass dontTouch my_inv_8x or // synopsys dont_touch
INVD8 my_inv_8x(.I(a), .ZN(b) );

```

```

module Count4(clk, reset, Q0, Q1, Q2, Q3); //1
input clk, reset; output Q0, Q1, Q2, Q3; wire Q0, Q1, Q2, Q3; //2
//           Q , D , clk, reset //3
asDff dff0( Q0, ~Q0, clk, reset); // The asDff is a //4
asDff dff1( Q1, ~Q1, Q0, reset); // standard component, //5
asDff dff2( Q2, ~Q2, Q1, reset); // unique to one set of tools. //6
asDff dff3( Q3, ~Q3, Q2, reset); //7
endmodule //8

```

```

module asDff (D, Q, Clk, Rst); //1
parameter width = 1, reset_value = 0; //2
input [width-1:0] D; output [width-1:0] Q; reg [width-1:0] Q; //3
input Clk, Rst; initial Q = {width{1'bx}}; //4
    always @ ( posedge Clk or negedge Rst ) //5

```

```

    if ( Rst==0 ) Q <= #1 reset_value;else Q <= #1 D;           //6
endmodule                                                       //7

```

12.5.12 Datapath Synthesis in Verilog

Key terms and concepts: Datapath synthesis • Synopsys VHDL DesignWare • compiler directives • X-BLOX • LPM (library of parameterized modules) • RPM (relationally placed modules) • thinking like the hardware

```

module DP_csum(A1,B1,Z1); input [3:0] A1,B1; output Z1; reg [3:0] Z1;
always@(A1 or B1) Z1 <= A1 + B1;//Compass adder_arch cond_sum_add
endmodule

```

```

module DP_ripp(A2,B2,Z2); input [3:0] A2,B2; output Z2; reg [3:0] Z2;
always@(A2 or B2) Z2 <= A2 + B2;//Compass adder_arch ripple_add
endmodule

```

```

module DP_sub_A(A,B,OutBus,CarryIn);                               //1
input [3:0] A, B ; input CarryIn ;                               //2
output OutBus ; reg [3:0] OutBus ;                               //3
always@(A or B or CarryIn) OutBus <= A - B - CarryIn ;         //4
endmodule                                                         //5

```

```

module DP_sub_B (A, B, CarryIn, Z) ;                             //1
input [3:0] A, B, CarryIn ; output [3:0] Z; reg [3:0] Z;       //2
always@(A or B or CarryIn) begin                                //3
    case (CarryIn)                                             //4
        1'b1 :      Z <= A - B - 1'b1;                        //5
        default : Z <= A - B - 1'b0; endcase                    //6
    end                                                         //7
endmodule                                                         //8

```

12.6 VHDL and Logic Synthesis

Key terms and concepts: IEEE VHDL nine-value system • You can use '1', 'H', '0', and 'L' in any manner • Some synthesis tools do not accept 'U' • You can use logic states 'Z', 'X', 'W', and '-' in assignments in any manner • 'Z' is synthesized to three-state logic • 'X', 'W',

and ' - ' are treated as unknown or don't care values • The IEEE synthesis packages provide the `STD_MATCH` function for comparisons

12.6.1 Initialization and Reset

Key terms and concepts: a VHDL `process` with a sensitivity list synthesizes to clocked logic with a reset

```
process (signal_1, signal_2) begin
  if (signal_2'EVENT and signal_2 = '0')
    then -- Insert initialization and reset statements.
    elsif (signal_1'EVENT and signal_1 = '1')
    then -- Insert clocking statements.
  end if;
end process;
```

12.6.2 Combinational Logic Synthesis in VHDL

Key terms and concepts: a **level-sensitive process** has a sensitivity list with signals that are not tested for event attributes ('EVENT or 'STABLE, for example) • combinational logic uses a level-sensitive `process` or a concurrent assignment statement • some synthesizers do not allow a signal inside a level-sensitive `process` unless the signal is in the sensitivity list

```
entity And_Bad is port (a, b: in BIT; c: out BIT); end And_Bad;
```

```
architecture Synthesis_Bad of And_Bad is
  begin process (a) -- this should be process (a, b)
    begin c <= a and b;
  end process;
end Synthesis_Bad;
```

12.6.3 Multiplexers in VHDL

Key terms and concepts: multiplexers can be synthesized using an (exhaustive) `case` statement (avoid the reserved word 'select') • a concurrent signal assignment is equivalent

```

entity Mux4 is port
(i: BIT_VECTOR(3 downto 0); sel: BIT_VECTOR(1 downto 0); s: out BIT);
end Mux4;

```

```

architecture Synthesis_1 of Mux4 is
  begin process(sel, i) begin
    case sel is
      when "00" => s <= i(0); when "01" => s <= i(1);
      when "10" => s <= i(2); when "11" => s <= i(3);
    end case;
  end process;
end Synthesis_1;

```

```

architecture Synthesis_2 of Mux4 is
  begin with sel select s <=
    i(0) when "00", i(1) when "01", i(2) when "10", i(3) when "11";
end Synthesis_2;

```

```

library IEEE; use ieee.std_logic_1164 all;
entity Mux8 is port
(InBus : in STD_LOGIC_VECTOR(7 downto 0);
Sel : in INTEGER range 0 to 7;
OutBit : out STD_LOGIC);
end Mux8;

```

```

architecture Synthesis_1 of Mux8 is
  begin process(InBus, Sel)
    begin OutBit <= InBus(Sel);
  end process;
end Synthesis_1;

```

12.6.4 Decoders in VHDL

```

library IEEE; --1
use IEEE.STD_LOGIC_1164 all; use IEEE.NUMERIC_STD all; --2
entity Decoder is port (enable : in BIT; --3
  Din: STD_LOGIC_VECTOR (2 downto 0); --4
  Dout: out STD_LOGIC_VECTOR (7 downto 0)); --5
end Decoder; --6

```

```

architecture Synthesis_1 of Decoder is                                --7
  begin                                                                --8
  with enable select Dout <=                                         --9
  STD_LOGIC_VECTOR                                                    --10
  (UNSIGNED'                                                           --11
    (shift_left                                                         --12
      ("00000001", TO_INTEGER (UNSIGNED(Din))                          --13
    )                                                                     --14
  )                                                                     --15
)                                                                        --16
  when '1',                                                            --17
  "11111111" when '0', "00000000" when others;                    --18
end Synthesis_1;                                                    --19

library IEEE;                                                        --1
use IEEE.NUMERIC_STD all; use IEEE.STD_LOGIC_1164 all;         --2

entity Concurrent_Decoder is port (                                  --3
  enable : in BIT;                                                    --4
  Din : in STD_LOGIC_VECTOR (2 downto 0);                            --5
  Dout : out STD_LOGIC_VECTOR (7 downto 0));                          --6
end Concurrent_Decoder;                                             --7

architecture Synthesis_1 of Concurrent_Decoder is                --8
begin process (Din, enable)                                          --9
  variable T : STD_LOGIC_VECTOR(7 downto 0);                        --10
  begin                                                                --11
  if (enable = '1') then                                            --12
    T := "00000000"; T( TO_INTEGER (UNSIGNED(Din))) := '1';         --13
    Dout <= T ;                                                       --14
  else Dout <= (others => 'Z');                                       --15
  end if;                                                            --16
end process;                                                         --17
end Synthesis_1;                                                    --18

```

12.6.5 Adders in VHDL

Key terms and concepts: To add two n -bit numbers and keep the overflow bit, we need to assign to a signal with more bits

```

library IEEE;                                                        --1
use IEEE.NUMERIC_STD all; use IEEE.STD_LOGIC_1164 all;         --2

entity Adder_1 is                                                  --3
port (A, B: in UNSIGNED(3 downto 0); C: out UNSIGNED(4 downto 0)); --4
end Adder_1;                                                         --5

architecture Synthesis_1 of Adder_1 is                            --6

```

```

    begin C <= ('0' & A) + ('0' & B);           --7
end Synthesis_1;                               --8

```

12.6.6 Sequential Logic in VHDL

Key terms and concepts: Sensitivity to an edge implies sequential logic in VHDL • Either: (1) no sensitivity list with a **wait until** statement (2) a sensitivity list and test for 'EVENT plus a specific level • any signal assigned in an edge-sensitive **process** statement should be reset—but be careful to distinguish between asynchronous and synchronous resets

```

library IEEE; use IEEE.STD_LOGIC_1164 all; entity DFF_With_Reset is
    port(D, Clk, Reset : in STD_LOGIC; Q : out STD_LOGIC);
end DFF_With_Reset;

```

```

architecture Synthesis_1 of DFF_With_Reset is
    begin process(Clk, Reset) begin
        if (Reset = '0') then Q <= '0'; -- asynchronous reset
            elsif rising_edge(Clk) then Q <= D;
        end if;
    end process;
end Synthesis_1;

```

```

architecture Synthesis_2 of DFF_With_Reset is
    begin process begin
        wait until rising_edge(Clk);
-- This reset is gated with the clock and is synchronous:
        if (Reset = '0') then Q <= '0'; else Q <= D; end if;
    end process;
end Synthesis_2;

```

Key terms and concepts: sequential logic results when we have to “remember” something between successive executions of a **process** statement. This occurs when a **process** statement contains one or more of the following situations (1) A signal is read but is not in the

sensitivity list of a **process** statement (2) A signal or variable is read before it is updated (3) A signal is not always updated (4) There are multiple **wait** statements

Not all of the models that we could write using the above constructs will be synthesizable. Any models that do use one or more of these constructs and that are synthesizable will result in sequential logic.

12.6.7 Instantiation in VHDL

Key terms and concepts: to help hand instantiate a component generate a structural netlist

```

`timescale 1ns/1ns //1
module halfgate (myInput, myOutput); //2
input myInput; output myOutput; wire myOutput; //3
    assign myOutput = ~myInput; //4
endmodule //5

library IEEE; use IEEE.STD_LOGIC_1164 all; --1
library COMPASS_LIB; use COMPASS_LIB.COMPASS all; --2
--compass compile_off -- synopsys etc. --3
use COMPASS_LIB.COMPASS_ETC all; --4
--compass compile_on -- synopsys etc. --5
entity halfgate_u is --6
--compass compile_off -- synopsys etc. --7
generic ( --8
    myOutput_cap : Real := 0.01; --9
    INSTANCE_NAME : string := "halfgate_u" ); --10
--compass compile_on -- synopsys etc. --11
port ( myInput : in Std_Logic := 'U'; --12
myOutput : out Std_Logic := 'U' ); --13
end halfgate_u; --14

architecture halfgate_u of halfgate_u is --15
component in01d0 --16
port ( I : in Std_Logic; ZN : out Std_Logic ); end component; --17
begin --18
u2: in01d0 port map ( I => myInput, ZN => myOutput ); --19
end halfgate_u; --20

--compass compile_off -- synopsys etc. --21
library cb60hd230d; --22
configuration halfgate_u_CON of halfgate_u is --23
    for halfgate_u --24
        for u2 : in01d0 use configuration cb60hd230d.in01d0_CON --25
            generic map ( --26

```

```

        ZN_cap => 0.0100 + myOutput_cap,           --27
        INSTANCE_NAME => INSTANCE_NAME&"/u2" )    --28
    port map ( I => I, ZN => ZN);                 --29
    end for;                                     --30
end for;                                         --31
end halfgate_u_CON;                             --32
--compass compile_on -- synopsys etc.          --33

component ASDFF
    generic (WIDTH : POSITIVE := 1;
        RESET_VALUE : STD_LOGIC_VECTOR := "0" );
    port      (Q      : out STD_LOGIC_VECTOR (WIDTH-1 downto 0);
        D      : in  STD_LOGIC_VECTOR (WIDTH-1 downto 0);
        CLK    : in  STD_LOGIC;
        RST    : in  STD_LOGIC );
end component;

library IEEE, COMPASS_LIB;                      --1
use IEEE.STD_LOGIC_1164 all; use COMPASS_LIB.STDCOMP all; --2
entity Ripple_4 is                              --3
    port (Trig, Reset: STD_LOGIC; QN0_5x:out STD_LOGIC; --4
        Q : inout STD_LOGIC_VECTOR(0 to 3)); --5
end Ripple_4; --6
architecture structure of Ripple_4 is --7
    signal QN : STD_LOGIC_VECTOR(0 to 3); --8
    component in01d1 --9
port ( I : in Std_Logic; ZN : out Std_Logic );end component; --10
    component in01d5 --11
port ( I : in Std_Logic; ZN : out Std_Logic );end component; --12
begin --13
--compass dontTouch inv5x -- synopsys dont_touch etc. --14
-- Named association for hand-instantiated library cells: --15
    inv5x: IN01D5 port map( I=>Q(0), ZN=>QN0_5x ); --16
    inv0 : IN01D1 port map( I=>Q(0), ZN=>QN(0) ); --17
    inv1 : IN01D1 port map( I=>Q(1), ZN=>QN(1) ); --18
    inv2 : IN01D1 port map( I=>Q(2), ZN=>QN(2) ); --19
    inv3 : IN01D1 port map( I=>Q(3), ZN=>QN(3) ); --20
-- Positional association for standard components: --21
--
--          Q          D          Clk  Rst          --22
d0: asDFF port map(Q (0 to 0), QN(0 to 0), Trig, Reset); --23
d1: asDFF port map(Q (1 to 1), QN(1 to 1), Q(0), Reset); --24
d2: asDFF port map(Q (2 to 2), QN(2 to 2), Q(1), Reset); --25

```



```

    d3: asDFF port map(Q (3 to 3), QN(3 to 3), Q(2), Reset);      --26
end structure;                                                --27

`timescale 1ns / 10ps                                         //1
module ripple_4_u (trig, reset, qn0_5x, q);                   //2
input trig; input reset; output qn0_5x; inout [3:0] q;       //3
wire [3:0] qn; supply1 VDD; supply0 VSS;                     //4
in01d5 inv5x (.I(q[0]),.ZN(qn0_5x));                           //5
in01d1 inv0 (.I(q[0]),.ZN(qn[0]));                             //6
in01d1 inv1 (.I(q[1]),.ZN(qn[1]));                             //7
in01d1 inv2 (.I(q[2]),.ZN(qn[2]));                             //8
in01d1 inv3 (.I(q[3]),.ZN(qn[3]));                             //9
dfctnb d0(.D(qn[0]),.CP(trig),.CDN(reset),.Q(q[0]),.QN(\d0.QN )); //10
dfctnb d1(.D(qn[1]),.CP(q[0]),.CDN(reset),.Q(q[1]),.QN(\d1.QN )); //11
dfctnb d2(.D(qn[2]),.CP(q[1]),.CDN(reset),.Q(q[2]),.QN(\d2.QN )); //12
dfctnb d3(.D(qn[3]),.CP(q[2]),.CDN(reset),.Q(q[3]),.QN(\d3.QN )); //13
endmodule                                                       //14

```

12.6.8 Shift Registers and Clocking in VHDL

```

library IEEE;                                                --1
use IEEE.STD_LOGIC_1164 all; use IEEE.NUMERIC_STD all;      --2

entity SIPO_1 is port (                                       --3
    Clk : in STD_LOGIC;                                       --4
    SI : in STD_LOGIC; -- serial in                            --5
    PO : buffer STD_LOGIC_VECTOR(3 downto 0)); -- parallel out --6
end SIPO_1;                                                  --7

architecture Synthesis_1 of SIPO_1 is                        --8
    begin process (Clk) begin                                 --9
        if (Clk = '1') then PO <= SI & PO(3 downto 1); end if; --10
    end process;                                             --11
end Synthesis_1;                                           --12

module sipo_1_u (clk, si, po);                                //1
input clk; input si; output [3:0] po;                        //2
supply1 VDD; supply0 VSS;                                    //3
dfntnb po_ff_b0 (.D(po[1]),.CP(clk),.Q(po[0]),.QN(\po_ff_b0.QN)); //4
dfntnb po_ff_b1 (.D(po[2]),.CP(clk),.Q(po[1]),.QN(\po_ff_b1.QN)); //5
dfntnb po_ff_b2 (.D(po[3]),.CP(clk),.Q(po[2]),.QN(\po_ff_b2.QN)); //6
dfntnb po_ff_b3 (.D(si),.CP(clk),.Q(po[3]),.QN(\po_ff_b3.QN )); //7
endmodule                                                    //8

library IEEE;                                                --1
use IEEE.STD_LOGIC_1164 all; use IEEE.NUMERIC_STD all;      --2

```

```

entity SIPO_R is port (                                     --3
  clk : in STD_LOGIC ; res : in STD_LOGIC ;                 --4
  SI : in STD_LOGIC ; PO : out STD_LOGIC_VECTOR(3 downto 0)); --5
end;                                                         --6

architecture Synthesis_1 of SIPO_R is                       --7
  signal PO_t : STD_LOGIC_VECTOR(3 downto 0);                --8
  begin                                                         --9
  process (PO_t) begin PO <= PO_t; end process;              --10
  process (clk, res) begin                                    --11
    if (res = '0') then PO_t <= (others => '0');              --12
    elsif (rising_edge(clk)) then PO_t <= SI & PO_t(3 downto 1); --13
    end if;                                                  --14
  end process;                                              --15
end Synthesis_1;                                           --16

```

12.6.9 Adders and Arithmetic Functions

Key terms and concepts: to perform BIT_VECTOR or STD_LOGIC_VECTOR arithmetic you have three choices: (1) Use a vendor-supplied package (2) Convert to SIGNED (or UNSIGNED) and use the IEEE standard synthesis packages (3) Use overloaded functions in packages or functions that you define yourself

```

library IEEE;                                               --1
use IEEE.STD_LOGIC_1164 all; use IEEE.NUMERIC_STD all; --2

entity Adder4 is port (                                     --3
  in1, in2 : in BIT_VECTOR(3 downto 0) ;                   --4
  mySum : out BIT_VECTOR(3 downto 0) ) ;                   --5
end Adder4;                                                 --6

architecture Behave_A of Adder4 is                         --7
  function DIY(L,R: BIT_VECTOR(3 downto 0)) return BIT_VECTOR is --8
  variable sum:BIT_VECTOR(3 downto 0);variable lt,rt,st,cry: BIT; --9
  begin cry := '0';                                         --10
  for i in L'REVERSE_RANGE loop                               --11
    lt := L(i); rt := R(i); st := ltxor rt;                 --12
    sum(i):= st xor cry; cry:= (lt and rt) or (st and cry); --13
  end loop;                                                --14
  return sum;                                               --15
  end;                                                     --16
  begin mySum <= DIY (in1, in2); -- do it yourself (DIY) add --17
end Behave_A;                                             --18

library IEEE;                                               --1
use IEEE.STD_LOGIC_1164 all; use IEEE.NUMERIC_STD all; --2

```

```

entity Adder4 is port ( --3
  in1, in2 : in UNSIGNED(3 downto 0) ; --4
  mySum : out UNSIGNED(3 downto 0) ) ; --5
end Adder4; --6

architecture Behave_B of Adder4 is --7
  begin mySum <= in1 + in2; -- This uses an overloaded '+'. --8
end Behave_B; --9

```

12.6.10 Adder/Subtractor and Don't Cares

Key terms and concepts: whether to use simple code or more complex code that more accurately describes the hardware?

```

library IEEE; --1
use IEEE.STD_LOGIC_1164 all; use IEEE.NUMERIC_STD all; --2
entity Adder_Subtractor is port ( --3
  xin : in UNSIGNED(15 downto 0); --4
  clk, addsub, clr: in STD_LOGIC; --5
  result : out UNSIGNED(15 downto 0)); --6
end Adder_Subtractor; --7

architecture Behave_A of Adder_Subtractor is --8
  signal addout, result_t: UNSIGNED(15downto 0); --9
  begin --10
    result <= result_t; --11
    with addsub select --12
    addout <= (xin + result_t)when '1', --13
              (xin - result_t)when '0', --14
              (others => '-') when others; --15
    process (clr, clk) begin --16
      if (clr = '0') then result_t <= (others => '0'); --17
      elsif rising_edge(clk) then result_t <= addout; --18
      end if; --19
    end process; --20
end Behave_A; --21

architecture Behave_B of Adder_Subtractor is --1
  signal result_t: UNSIGNED(15 downto 0); --2
  begin --3
    result <= result_t; --4
    process (clr, clk) begin --5
      if (clr = '0') then result_t <= (others => '0'); --6
      elsif rising_edge(clk) then --7
        case addsub is --8
          when '1' => result_t <= (xin + result_t); --9
          when '0' => result_t <= (xin - result_t); --10

```



```

end //19
always @(curSt or yOut) // Assign the next state: //20
begin:Comb //21
    case (curSt) //22
        `resSt:nextSt = `S3; `S1:nextSt = `S2; //23
        `S2:nextSt = `S1; `S3:nextSt = `S1; //24
        default:nextSt = `resSt; //25
    endcase //26
end //27
endmodule //28

module StateMachine_2 (reset, clk, yOutReg); //1
    input reset, clk; output yOutReg; reg yOutReg, yOut; //2
    parameter [1:0] //synopsys enum states //3
        resSt = 2'b00, S1 = 2'b01, S2 = 2'b10, S3 = 2'b11; //4
    reg [1:0] /* synopsys enum states */ curSt, nextSt; //5
//synopsys state_vector curSt //6
always @(posedge clk or posedge reset) begin //7
    if (reset == 1) //8
        begin yOut = 0; yOutReg = yOut; curSt = resSt;end //9
    else begin //10
        case (curSt) resSt:yOut = 0;S1:yOut = 1;S2:yOut = 1;S3:yOut = 1; //11
            default:yOut = 0; endcase //12
        yOutReg = yOut; curSt = nextSt;end //13
    end //14
always @(curSt or yOut) begin //15
    case (curSt) //16
        resSt:nextSt = S3; S1:nextSt = S2; S2:nextSt = S1; S3:nextSt = S1; //17
        default:nextSt = S1; endcase //18
    end //19
endmodule //20

parameter [3:0] //synopsys enum states
    resSt = 4'b0000, S1 = 4'b0010, S2 = 4'b0100, S3 = 4'b1000;

```

12.7.2 FSM Synthesis in VHDL

Key terms and concepts: Moore state machine • Mealy state machine • An FSM compiler extracts a state machine

```

library IEEE; use IEEE.STD_LOGIC_1164 all;                                --1
entity SM1 is                                                                --2
  port (aIn, clk : in Std_logic; yOut: out Std_logic);                       --3
end SM1;                                                                        --4

architecture Moore of SM1 is                                                --5
  type state is (s1, s2, s3, s4);                                           --6
  signal pS, nS : state;                                                     --7
  begin                                                                           --8
  process (aIn, pS) begin                                                    --9
    case pS is                                                                --10
    when s1 => yOut <= '0'; nS <= s4;                                         --11
    when s2 => yOut <= '1'; nS <= s3;                                         --12
    when s3 => yOut <= '1'; nS <= s1;                                         --13
    when s4 => yOut <= '1'; nS <= s2;                                         --14
    end case;                                                                  --15
  end process;                                                                --16
  process begin                                                                --17
    -- synopsys etc.                                                           --18
    --compass Statemachine adj pS                                             --19
    wait until clk = '1'; pS <= nS;                                          --20
  end process;                                                                --21
end Moore;                                                                      --22

library IEEE; use IEEE.STD_LOGIC_1164 all;                                --1
entity SM2 is                                                                --2
  port (aIn, clk : in Std_logic; yOut: out Std_logic);                       --3
end SM2;                                                                        --4

architecture Mealy of SM2 is                                                --1
  type state is (s1, s2, s3, s4);                                           --2
  signal pS, nS : state;                                                     --3
  begin                                                                           --4
  process(aIn, pS) begin                                                    --5
  case pS is                                                                --6
  when s1 => if (aIn = '1')                                                  --7
    then yOut <= '0'; nS <= s4;                                             --8
    else yOut <= '1'; nS <= s3;                                             --9
    end if;                                                                    --10
  when s2 => yOut <= '1'; nS <= s3;                                         --11
  when s3 => yOut <= '1'; nS <= s1;                                         --12

```

```

when s4 => if (aIn = '1')                                --13
    then yOut <= '1'; nS <= s2;                          --14
    else yOut <= '0'; nS <= s1;                          --15
    end if;                                              --16
end case;                                              --17
end process;                                          --18
process begin                                          --19
wait until clk = '1' ;                                --20
--Compass Statemachine oneHot pS                      --21
pS <= nS;                                           --22
end process;                                          --23
end Mealy;                                           --24

```

12.8 Memory Synthesis

Key terms and concepts: approaches to memory synthesis: (1) Random logic using flip-flops or latches (2) Register files in datapaths (3) RAM standard components (4) RAM compilers

12.8.1 Memory Synthesis in Verilog

Key terms and concepts: Verilog memory array • an array of latches or flip-flops

```

reg [31:0] MyMemory [3:0]; // a 4 x 32-bit register

module RAM_1(A, CEB, WEB, OEB, INN, OUTT);              //1
    input [6:0] A; input CEB,WEB,OEB; input [4:0]INN;   //2
    output [4:0] OUTT;                                  //3
    reg [4:0] OUTT; reg [4:0] int_bus; reg [4:0] memory [127:0]; //4
always@(negedge CEB) begin                             //5
    if (CEB == 0) begin                                 //6
        if (WEB == 1) int_bus = memory[A];             //7
        else if (WEB == 0) begin memory[A] = INN; int_bus = INN;end //8
        else int_bus = 5'bxxxxx;                      //9
    end                                                 //10
end                                                     //11
always@(OEB or int_bus) begin                          //12
    case (OEB) 0 : OUTT = int_bus;                     //13
        default : OUTT = 5'bzzzzz; endcase             //14

```

```

end //15
endmodule //16

memory[i + 1] = memory[i]; // needs two clock cycles
pointer = memory[memory[i]]; // needs two clock cycles
pc = memory[addr1]; memory[addr2] = pc + 1; // not on the same cycle

```

12.8.2 Memory Synthesis in VHDL

Key terms and concepts: VHDL multidimensional arrays • array of latches • standard-cell RAM

```

type memStor is array(3 downto 0) of integer; -- This is OK.

subtype MemReg is STD_LOGIC_VECTOR(15 downto 0); -- So is this.
type memStor is array(3 downto 0) of MemReg;
-- other code...
signal Mem1 : memStor;

library IEEE; --1
use IEEE.STD_LOGIC_1164 all; --2
package RAM_package is --3
constant numOut : INTEGER := 8; --4
constant wordDepth: INTEGER := 8; --5
constant numAddr : INTEGER := 3; --6
subtype MEMV is STD_LOGIC_VECTOR(numOut-1 downto 0); --7
type MEM is array (wordDepth-1 downto 0) of MEMV; --8
end RAM_package; --9

library IEEE; --10
use IEEE.STD_LOGIC_1164 all; use IEEE.NUMERIC_STD all; --11
use work.RAM_package all; --12
entity RAM_1 is --13
  port (signal A : in STD_LOGIC_VECTOR(numAddr-1 downto 0); --14
    signal CEB, WEB, OEB : in STD_LOGIC; --15
    signal INN : in MEMV; --16
    signal OUTT : out MEMV); --17
end RAM_1; --18

architecture Synthesis_1 of RAM_1 is --19
  signal i_bus : MEMV; -- RAM internal data latch --20
  signal mem : MEM; -- RAM data --21
  begin --22
  process begin --23
    wait until CEB = '0'; --24

```



```

    if WEB = '1' then i_bus <= mem(TO_INTEGER(UNSIGNED(A))); --25
    elsif WEB = '0' then --26
        mem(TO_INTEGER(UNSIGNED(A))) <= INN; --27
        i_bus <= INN; --28
    else i_bus <= (others => 'X'); --29
    end if; --30
end process; --31

process(OEB, int_bus) begin -- control output drivers: --32
    case (OEB) is --33
        when '0' => OUTT <= i_bus; --34
        when '1' => OUTT <= (others => 'Z'); --35
        when others => OUTT <= (others => 'X'); --36
    end case; --37
end process; --38
end Synthesis_1; --39

```

12.9 The Multiplier

Key terms and concepts: warnings and errors during elaboration

```
Sum <= X xor Y xor Cin after TS;
```

Warning: AFTER clause in a waveform element is not supported

```
port (A, B : in BIT_VECTOR (7 downto 0); Sel : in BIT := '0'; Y : out
BIT_VECTOR (7 downto 0));
```

Warning: Default values on interface signals are not supported

```
port (X:BIT_VECTOR; F:out BIT );
```

Error: An index range must be specified for this data type

```
begin assert (D'LENGTH <= Q'LENGTH)
    report "D wider than output Q" severity Failure;
```

Warning: Assertion statements are ignored

Error: Statements in entity declarations are not supported

```
if CLR = '1' then St := (others => '0'); Q <= St after TCQ;
```

Error: Illegal use of aggregate with the choice "others": the derived subtype of an array aggregate that has a choice "others" must be a constrained array subtype

```
signal SRA, SRB, ADDout, MUXout, REGout: BIT_VECTOR(7 downto 0);
```

Warning: Name is reserved word in VHDL-93: sra

```
signal Zero, Init, Shift, Add, Low: BIT := '0'; signal High: BIT := '1';
```

Warning: Initial values on signals are only for simulation and setting the value of undriven signals in synthesis. A synthesized circuit can not be guaranteed to be in any known state when the power is turned on.

12.9.1 Messages During Synthesis

Key terms and concepts: error and warning messages during synthesis

These unused instances are being removed: in full_adder_p_dup8: u5, u2, u3, u4

These unused instances are being removed: in dffclr_p_dup1: u2

```
architecture Behave of DFFClr is --1
  signal Qi : BIT; --2
  begin QB <= not Qi; Q <= Qi; --3
  process (CLR, CLK) begin --4
    if CLR = '1' then Qi <= '0' after TRQ; --5
    elsif CLK'EVENT and CLK = '1' then Qi <= D after TCQ; --6
    end if; --7
  end process; --8
end; --9
```

```
A1: Adder8 port map(A=>SRB,B=>REGout,Cin=>Low,Cout=>OFL,Sum=>ADDout);
```

```
Cout <= (X and Y) or (X and Cin) or (Y and Cin) after TC;
```

12.10 The Engine Controller

Key terms and concepts: warnings and errors during optimization • unassigned or uninitialized variables

Warning: Made latches to store values on: net d(4), d(5), d(6), d(7), d(8), d(9), d(10), d(11), in module fifo_control

```

case sel is
  when "01" => D <= D_1 after TPD; r1 <= '1' after TPD;
  when "10" => D <= D_2 after TPD; r2 <= '1' after TPD;
  when "00" => D(3) <= f1 after TPD; D(2) <= f2 after TPD;
                D(1) <= e1 after TPD; D(0) <= e2 after TPD; -- Bad!
  when others => D <= "ZZZZZZZZZZZZ" after TPD;
end case;

  when "00" => D(3) <= f1 after TPD; D(2) <= f2 after TPD; -- Write
                D(1) <= e1 after TPD; D(0) <= e2 after TPD; -- to
                D(11 downto 4) <= "ZZZZZZZZ" after TPD; -- all bits.

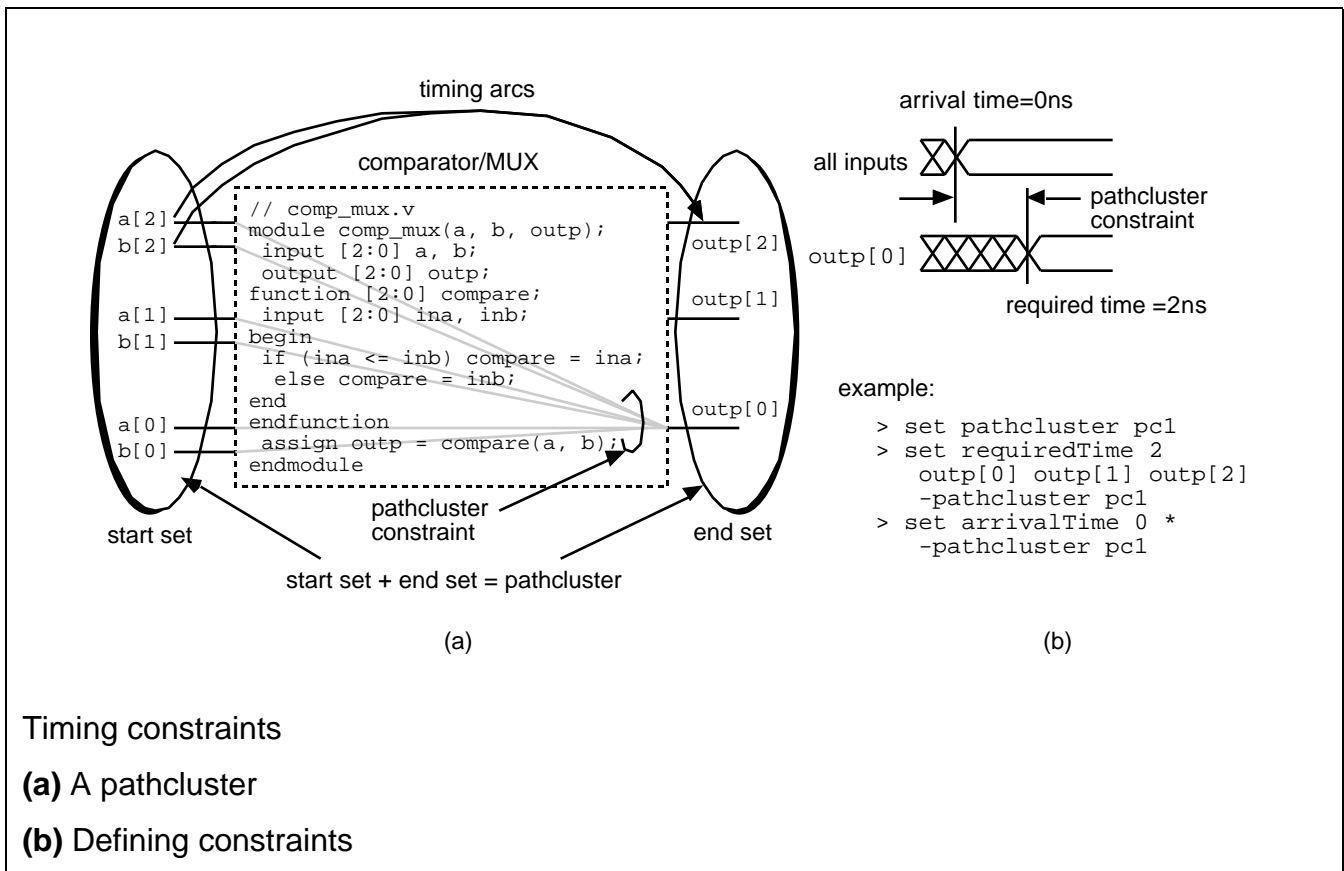
```

12.11 Performance-Driven Synthesis

Key terms and concepts: use of directives and pseudocomments • **timing arcs** (or timing paths)
 • a **pathcluster** (a group of circuit nodes) • **required time** for a signal to reach the output nodes (the **end set**) • **arrival time** of the signals at all the inputs • constrained delay • timing constraint
 • **slack** • the timing constraint is **met** or **violated**

12.12 Optimization of the Viterbi Decoder

Key terms and concepts: set the **environment** using worst-case conditions • die temperature of 25°C (fastest logic) to 120°C (slowest logic) • power supply voltage of $V_{DD}=5.5V$ (fastest logic) to $V_{DD}=4.5V$ (slowest logic) • worst process (slowest logic) to best process (fastest logic)



12.13 Summary

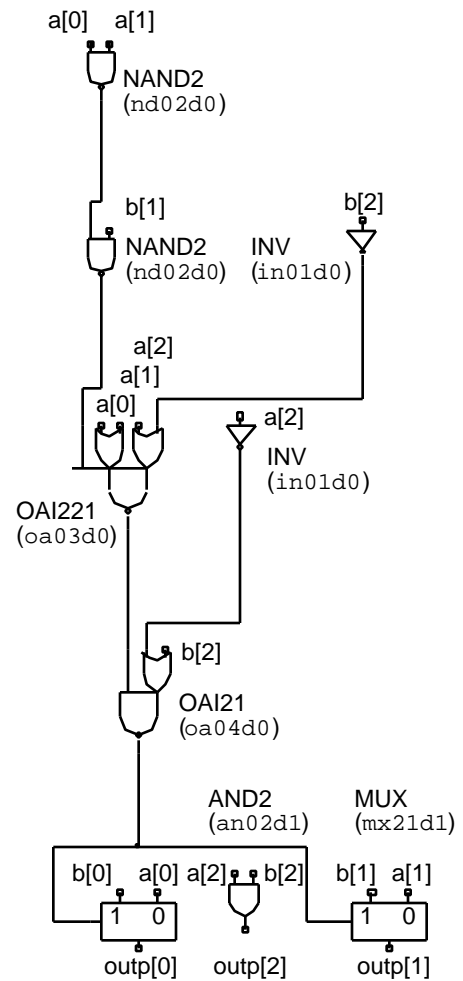
Key terms and concepts: A logic synthesizer may contain over 500,000 lines of code • danger of the “garbage in, garbage out” syndrome • “What do I expect to see at the output?” • “Does the output make sense?” • the worst thing you can do is write and simulate a huge amount of code, read it into the synthesis tool, and try and optimize it all at once with the default settings • interconnect delay is increasingly dominant • it is important to begin physical design as early as possible • ideally floorplanning and logic synthesis should be completed at the same time

```

`timescale 1ns / 10ps
module comp_mux_o (a, b, outp);
input [2:0] a; input [2:0] b;
output [2:0] outp;
supply1 VDD; supply0 VSS;

mx21d1 B1_i1 (.I0(a[0]), .I1(b[0]),
.S(B1_i6_ZN), .Z(outp[0]));
oa03d1 B1_i2 (.A1(B1_i9_ZN), .A2(a[2]),
.B1(a[0]), .B2(a[1]), .C(B1_i4_ZN),
.ZN(B1_i2_ZN));
nd02d0 B1_i3 (.A1(a[1]), .A2(a[0]),
.ZN(B1_i3_ZN));
nd02d0 B1_i4 (.A1(b[1]), .A2(B1_i3_ZN),
.ZN(B1_i4_ZN));
mx21d1 B1_i5 (.I0(a[1]), .I1(b[1]),
.S(B1_i6_ZN), .Z(outp[1]));
oa04d1 B1_i6 (.A1(b[2]), .A2(B1_i7_ZN),
.B(B1_i2_ZN), .ZN(B1_i6_ZN));
in01d0 B1_i7 (.I(a[2]), .ZN(B1_i7_ZN));
an02d1 B1_i8 (.A1(b[2]), .A2(a[2]),
.Z(outp[2]));
in01d0 B1_i9 (.I(b[2]), .ZN(B1_i9_ZN));
endmodule

```



The comparator/MUX example after logic optimization with timing constraints

The structural netlist, `comp_mux_o2.v`, and its derived schematic

